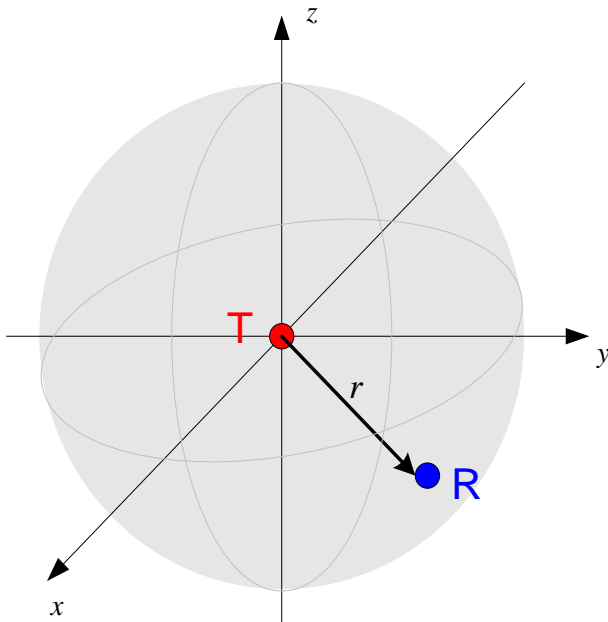


Lecture 2. Fading Channel

- Characteristics of Fading Channels
- Modeling of Fading Channels
- Discrete-time Input/Output Model

Radio Propagation in Free Space

- Speed: $c = 299,792,458$ m/s
- Isotropic
 - Received power at a particular location decays with distance:
 $P \sim r^{-2}$



✓ Transmitted signal: $s(t) = \cos 2\pi ft$

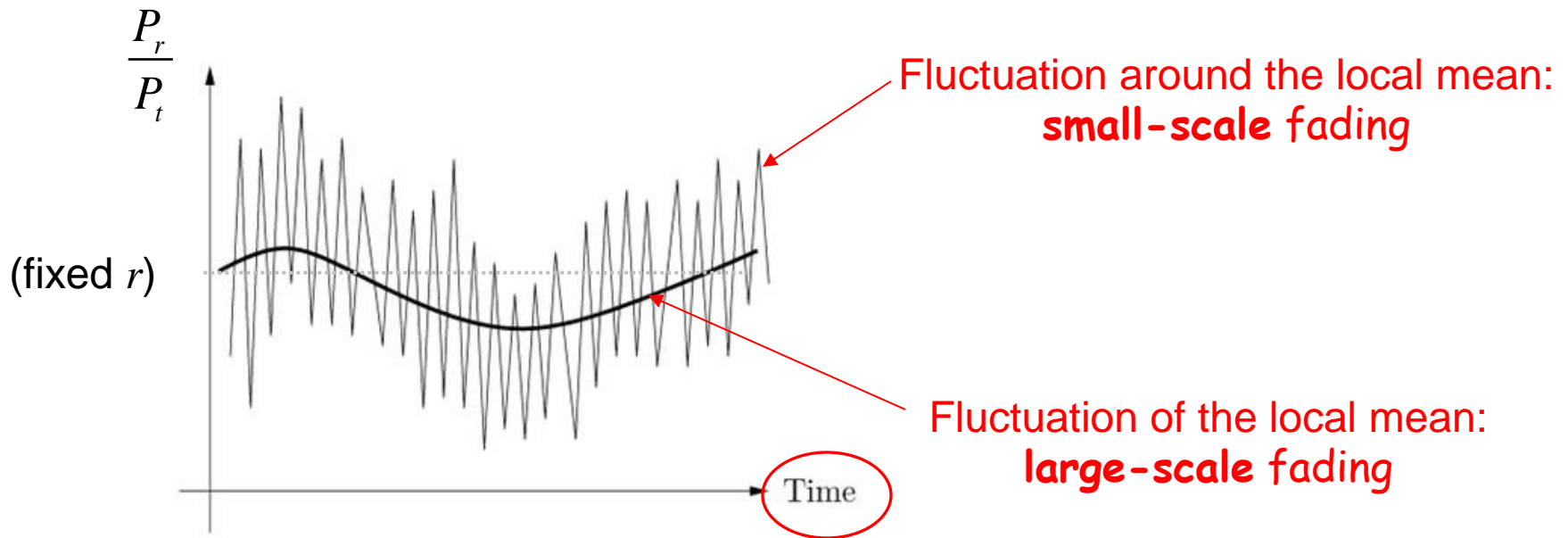
✓ Received signal: $y(t) = \frac{\beta \cos 2\pi f(t - r/c)}{r}$

Radio Propagation in Terrestrial Environment



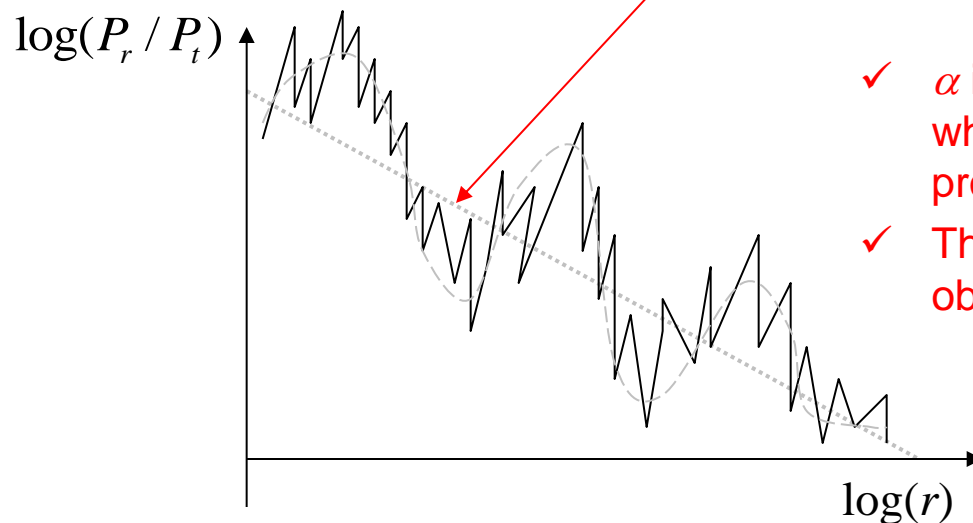
Transmission Power: P_t

Received Power: P_r



Large-scale Fading

- Large-scale fading -- Log-normal shadowing
 - Attributed to the random variation of propagation environment.
 - Empirically modeled as a log-normal random variable with mean μ and variance σ^2 .
 - Mean μ is determined by the path loss: $\mu \sim r^{-\alpha}$, $\alpha > 2$.



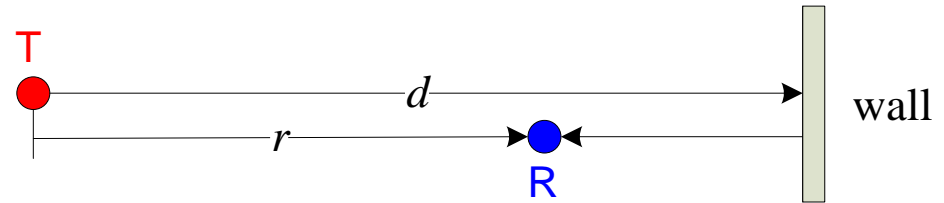
- ✓ α is called path-loss factor, which depends on the propagation environment.
- ✓ The higher the density of obstacles, the larger α .

Small-scale Fading

- Small-scale fading
 - Attributed to 1) multiple arriving paths; and 2) movements of the transmitter and/or receiver.

A Simple Two-path Model (1)

- Reflecting wall, fixed antenna



✓ Transmitted signal: $s(t) = \cos 2\pi ft$

✓ Received signal: $y(t) = \frac{\beta \cos 2\pi f(t - r/c)}{r} - \frac{\beta \cos 2\pi f(t - (2d - r)/c)}{2d - r}$

The phase difference between the two waves:

$$\begin{aligned}\Delta\phi &= \left(\frac{2\pi f(2d - r)}{c} + \pi \right) - \left(\frac{2\pi fr}{c} \right) \\ &= \frac{4\pi f}{c}(d - r) + \pi\end{aligned}$$

What happens if $\Delta\phi$ changes from π to 2π ?

$\Delta\phi = 2k\pi$: Constructively add

$\Delta\phi = (2k+1)\pi$: Destructively add

Coherence Bandwidth vs. Delay Spread

- An appreciable change of received signal will be observed if the **frequency** of transmitted signal f changes by:

$$\frac{c}{4(d-r)}$$

Coherence
Bandwidth
 W_c

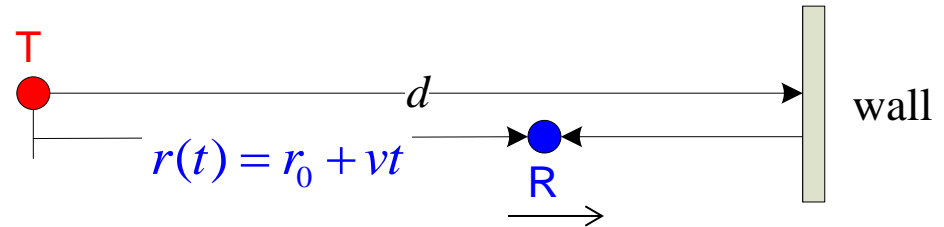
- The difference between the propagation **delays** along the two signal paths is: $\frac{2d-r}{c} - \frac{r}{c} = \frac{2(d-r)}{c}$

Delay
Spread
 T_d

- $T_d \sim \frac{1}{W_c}$

A Simple Two-path Model (2)

- Reflecting wall, moving antenna



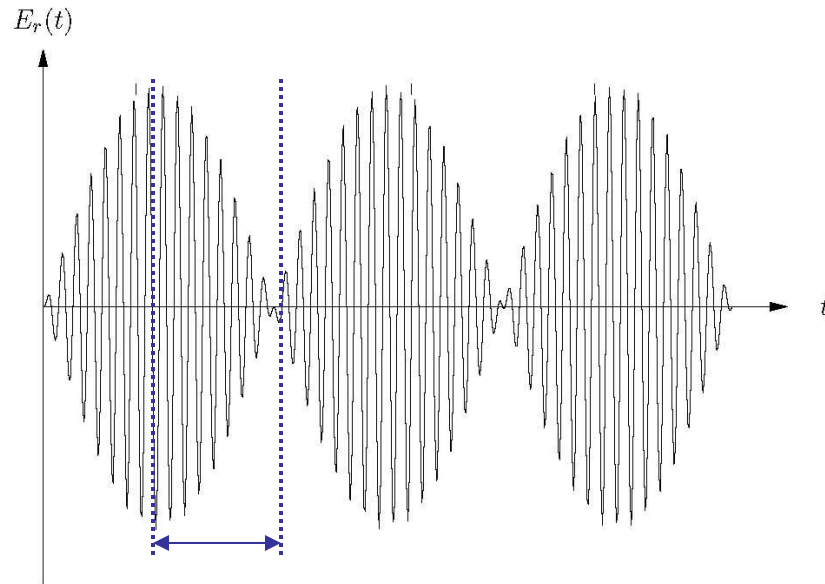
✓ Transmitted signal: $s(t) = \cos 2\pi ft$

✓ Received signal:

$$\begin{aligned}
 y(t) &= \frac{\beta \cos 2\pi f(t - r(t)/c)}{r(t)} - \frac{\beta \cos 2\pi f(t - (2d - r(t))/c)}{2d - r(t)} \\
 &\stackrel{\text{Doppler Shift: } D_1 = -fv/c}{=} \frac{\beta \cos 2\pi f((1 - v/c)t - r_0/c)}{r_0 + vt} - \frac{\beta \cos 2\pi f((1 + v/c)t + (r_0 - 2d)/c)}{2d - r_0 - vt} \\
 &\approx \frac{2\beta \sin 2\pi f(vt/c + (r_0 - d)/c) \sin 2\pi f(t - d/c)}{r_0 + vt}
 \end{aligned}$$

Coherence Time vs. Doppler Spread

- An appreciable change of received signal will be observed if time t changes by: $\frac{c}{4fv}$



- The difference between the Doppler shifts of the two paths is:

$$D_s = D_2 - D_1 = 2fv/c$$

- $T_c \sim \frac{1}{D_s}$



Modeling of Fading Channels



- A fading channel can be modeled as a **Linear Time-Varying** system.

$$y(t) = \int_{-\infty}^{\infty} h(\tau, t) s(t - \tau) d\tau$$

- Suppose that $h(\tau, t)$ is a deterministic function of time t and delay τ .

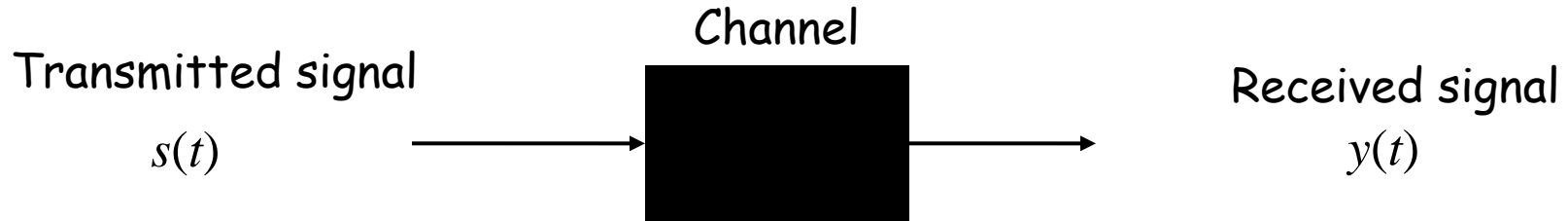
Time-variant transfer function: $H(f, t) = \mathcal{F}_{\tau}(h(\tau, t))$

Delay Doppler function: $S(\tau, \nu) = \mathcal{F}_t(h(\tau, t))$

Doppler-variant transfer function: $D(f, \nu) = \mathcal{F}_{\tau}(S(\tau, \nu)) = \mathcal{F}_t(H(f, t))$

$\mathcal{F}(\cdot)$ denotes Fourier transform.

Modeling of Fading Channels



- A fading channel can be modeled as a **Linear Time-Varying** system.

$$y(t) = \int_{-\infty}^{\infty} h(\tau, t) s(t - \tau) d\tau$$

- Suppose that $h(\tau, t)$ is a WSS process with uncorrelated scatterers:
The autocorrelation function $R_h(\tau_1, t_1, \tau_2, t_2) = E[h(\tau_1, t_1)h^*(\tau_2, t_2)] = R_h(\tau, \Delta t)$

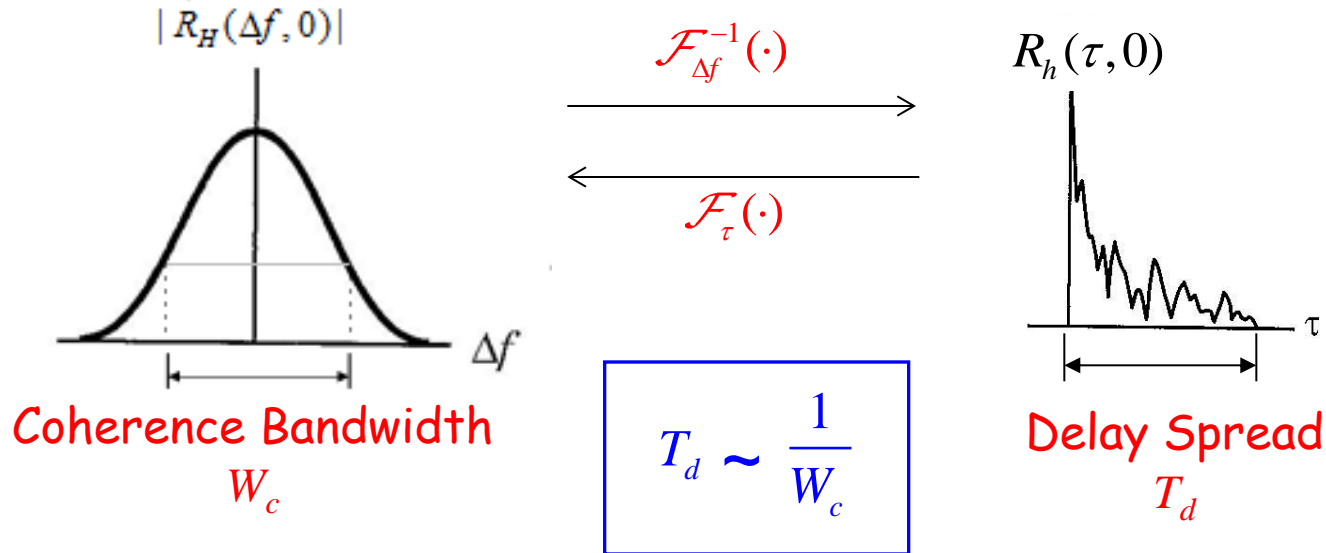
Time-variant transfer function: $R_H(\Delta f, \Delta t) = \mathcal{F}_{\tau}(R_h(\tau, \Delta t))$

Delay Doppler function: $R_S(\tau, \nu) = \mathcal{F}_{\Delta t}(R_h(\tau, \Delta t))$

Doppler-variant transfer function: $R_D(\Delta f, \nu) = \mathcal{F}_{\tau}(R_S(\tau, \nu)) = \mathcal{F}_{\Delta t}(R_H(\Delta f, \Delta t))$

Coherence Bandwidth vs. Delay Spread

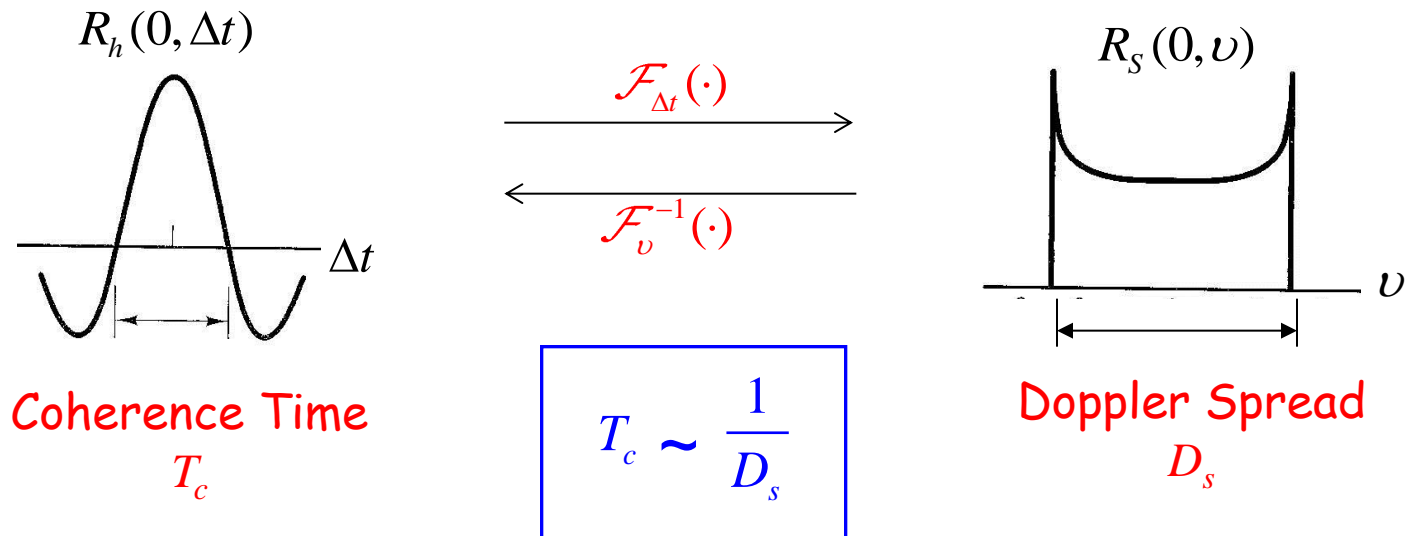
- Let $\Delta t = 0$:



- Flat fading: Signal Bandwidth $W \ll$ Coherence Bandwidth W_c
Symbol Time $T \gg$ Delay Spread T_d
- Frequency-selective fading: Signal Bandwidth $W \gg$ Coherence Bandwidth W_c
Symbol Time $T \ll$ Delay Spread T_d

Coherence Time vs. Doppler Spread

- Let $\tau = 0$:



- **Slow fading:** Symbol Time $T \ll$ Coherence Time T_c
 Signal Bandwidth $W \gg$ Doppler Spread D_s
- **Fast fading:** Symbol Time $T \gg$ Coherence Time T_c
 Signal Bandwidth $W \ll$ Doppler Spread D_s

Discrete-time Input/Output Model

- $$y(t) = \int_{-\infty}^{\infty} h(\tau, t) s(t - \tau) d\tau = \sum_n s[n] \sum_i a_i(t) \Psi_{T_0}(t - \tau_i(t) - nT_0)$$

$$h(\tau, t) = \sum_i a_i(t) \delta(\tau - \tau_i(t))$$

$$s(t) = \sum_n s[n] \Psi_{T_0}(t - nT_0)$$

- $a_i(t)$: attenuation at time t of path i

- $\Psi_{T_0}(t)$: modulation pulse

- $\tau_i(t)$: propagation delay at time t of path i

- Sampled output at $t=mT_0$:

$$y[m] = \sum_n s[n] \sum_i a_i(mT_0) \Psi_{T_0}((m-n)T_0 - \tau_i(mT_0))$$

Let $l=m-n$.
$$y[m] = \sum_l s[m-l] \sum_i a_i(mT_0) \Psi_{T_0}(lT_0 - \tau_i(mT_0)) \quad h_l[m]$$

- Discrete-time Input/Output Model:
$$y[m] = \sum_l h_l[m] s[m-l] + z[m]$$

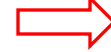
Discrete-time Input/Output Model

$$y[m] = \sum_{l=0}^{L-1} h_l[m]s[m-l] + z[m]$$

- How many channel filter taps can be obtained (how to determine the value of L)?

Delay spread: $T_d = \max_{i,j} |\tau_i - \tau_j|$

Sampling rate: W



$$L = \lceil WT_d \rceil$$

- Flat fading: $L=1$
 - Frequency-selective fading: $L \gg 1$
- How fast does the channel filter tap gain $h_l[m]$ vary with time in one symbol time T ? Depends on the coherence time T_c .
- Slow fading: $h_l[m]$ can be regarded as constant in T
 - Fast fading: $h_l[m]$ varies with time in T

More about $h_l[m]$

- Without line-of-sight (LOS) paths: $h_l[m]$ can be modeled as a zero-mean complex Gaussian random variable $\mathcal{CN}(0, \sigma_s^2)$.

➤ The magnitude follows Rayleigh distribution.

- With one LOS path: $h_l[m]$ can be modeled as

$$h_l[m] = \sqrt{\frac{k}{k+1}} \sigma_s e^{j\theta} + \sqrt{\frac{1}{k+1}} \mathcal{CN}(0, \sigma_s^2)$$


k : the ratio of the power in the LOS path and that in the scattered paths.

➤ The magnitude follows Ricean distribution.

More about $h_l[m]$

- Availability of Channel State Information (CSI)
 - CSIR: CSI is available at the receiver side.
 - available through channel measurement
 - required for coherent detection
 - CSIT: CSI is available at the transmitter side.
 - available through channel measurement and feedback
 - optional
- Ergodicity
 - indispensable condition to achieve the Shannon capacity
 - may not hold in delay-limited scenarios

Summary

- Fading Channel
 - Large-scale fading
 - Path loss
 - Shadowing
 - Small-scale fading
 - Delay spread (caused by multipath) 
 - Doppler spread (caused by mobility) 
 - Discrete-time input/output model

$$y[m] = \sum_{l=0}^{L-1} h_l[m]s[m-l] + z[m]$$